

AN116

**Buck-Boost Isolated Discontinuous Current Design
Using an LPT E2000Q Core**

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The Buck-Boost Isolated Converter is shown in Figure 1.

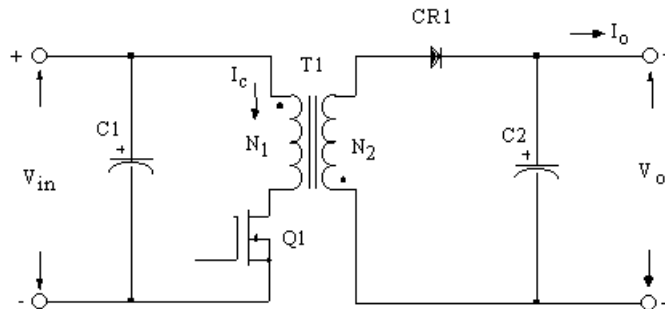


Figure 1. Buck-Boost isolated discontinuous current converter.

Buck-Boost Isolated Discontinuous Current Design Specification

1.	Input voltage nominal	$V_{nom} = 28 \text{ V}$
2.	Input voltage minimum	$V_{min} = 24 \text{ V}$
3.	Input voltage maximum	$V_{max} = 32 \text{ V}$
4.	Output voltage	$V_o = 5 \text{ V}$
5.	Output current	$I_o = 5 \text{ A}$
6.	Window utilization	$K_U = 0.4$
7.	Frequency	$f = 100 \text{ kHz}$
8.	Converter efficiency	$\eta = 90 \%$
9.	Maximum duty ratio	$D_{max} = 0.5$
10.	Dwell time duty ratio	$D_w = 0.1$
11.	Regulation	$\alpha = 1.0 \%$
12.	Operating flux density	$\Delta B = 0.4 \text{ T}$
13.	Diode voltage drop	$V_d = 1 \text{ V}$

At this point, select a wire so that the relationship between the ac resistance and the dc resistance is 1:

$$\frac{R_{AC}}{R_{DC}} = 1$$

The skin depth in cm is:

$$d = \frac{6.62}{\sqrt{f}}$$

$$d = \frac{6.62}{\sqrt{100,000}} = 0.0209 \text{ [cm]}$$

Then, the wire diameter is:

$$\text{Wire diameter} = 2d$$

$$\text{Wire diameter} = 2 \cdot 0.0209 = 0.0418 \text{ [cm]}$$

Then, the bare wire area A_w is:

$$A_w = \frac{\pi D^2}{4}$$

$$A_w = \frac{3.1416 \cdot 0.0418^2}{4} = 0.00137 \text{ [cm}^2\text{]}$$

From the Wire Table, number 26 has a bare wire area of 0.001280 cm². This will be the minimum wire size used in this design. If the design requires more wire area to meet the specification, then, the design will use a multifilar of #26. Listed below are #27 and #28, just in case #26 requires too much rounding off.

Wire AWG	Bare Area	Area Ins.	Bare/Ins.	$\mu\Omega/\text{cm}$
#26	0.00128	0.001603	0.798	1345
#27	0.001021	0.001313	0.778	1687
#28	0.000804	0.000105	0.765	2142

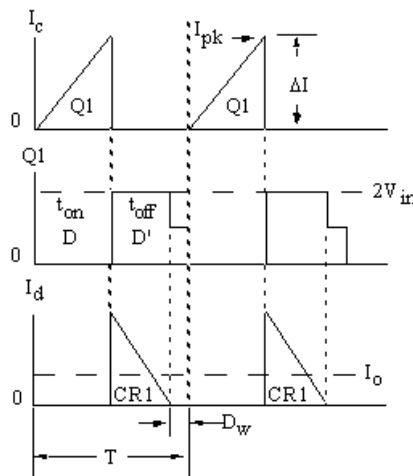


Figure 2. Voltage, current waveforms of a buck-boost isolated discontinuous current converter.

Step No. 1 Calculate the total period, T.

$$T = \frac{1}{f}$$

$$T = \frac{1}{100,000} = 10 \cdot 10^{-6} \text{ [s]}$$

Step No. 2 Calculate the maximum transistor on time, t_{on} .

$$t_{on} = TD_{MAX}$$

$$t_{on} = 10 \cdot 10^{-6} \cdot 0.5 = 5 \text{ [ms]}$$

Step No. 3 Calculate the secondary load power, P_o .

$$P_o = I_o (V_o + V_d)$$

$$P_o = 5 \cdot (5 + 1) = 30 \text{ [W]}$$

Step No. 4 Calculate the maximum input current, I_{\max} .

$$I_{in(max)} = \frac{P_o}{V_{min} \cdot h}$$

$$I_{in(max)} = \frac{30}{24 \cdot 0.9} = 1.39 \text{ [A]}$$

Step No. 5 Calculate the primary peak current, I_{ppk} .

$$I_{ppk} = \frac{2 \cdot T \cdot P_o}{h \cdot V_{in(min)} \cdot t_{on(max)}}$$

$$I_{ppk} = \frac{2 \cdot 10 \cdot 10^{-6} \cdot 30}{0.9 \cdot 24 \cdot 5 \cdot 10^{-6}} = 5.55 \text{ [A]}$$

Step No. 6 Calculate the primary rms current, I_{prms} .

$$I_{prms} = I_{ppk} \cdot \sqrt{\frac{t_{on}}{3 \cdot T}}$$

$$I_{prms} = 5.55 \cdot \sqrt{\frac{5}{3 \cdot 10}} = 2.27 \text{ [A]}$$

Step No. 7 Calculate the required primary inductance, L .

$$L = \frac{V_{in(min)} \cdot t_{on(max)}}{I_{ppk}}$$

$$L = \frac{24 \cdot 5 \cdot 10^{-6}}{5.55} = 21.6 \text{ [mH]}$$

Step No. 8 Calculate the energy-handling capability in watt-second.

$$ENG = \frac{L \cdot I_{ppk}^2}{2}$$

$$ENG = \frac{21.6 \cdot 10^{-6} \cdot 5.55^2}{2} = 0.000333 \text{ [W} \cdot \text{s]}$$

Step No. 9 Calculate the electrical conditions, K_e .

$$K_e = 0.145 \cdot P_o \cdot B_m^2 \cdot 10^{-4}$$

$$K_e = 0.145 \cdot 30 \cdot 0.4^2 \cdot 10^{-4} = 0.0000696$$

Step No. 10 Calculate the core geometry, K_g .

$$K_g = \frac{(ENERGY)^2}{K_e \cdot a}$$

$$K_g = \frac{0.000333^2}{0.0000696 \cdot 1.0} = 0.00159 \text{ [cm}^5\text{]}$$

Step No. 11. Select from the LPT data sheet a E2000Q core comparable in core geometry, K_g .

Core number	GC70111
Manufacturer	CMI
Magnetic path length, MPL	4.1 cm
Core weight, W_{tfe}	4.3 g
Copper weight, W_{tcu}	5.6 g
Mean length turn, MLT	2.7 cm
Iron area, A_c	0.14 cm ²
Window area, W_a	0.581 cm ²

Area product, A_p	0.08132 cm ⁴
Core geometry, K_g	0.00168 cm ⁵
Surface area, A_t	16.3 cm ²
Permeability, μ	300
MilliHenries per 1000 turns, mH	129

Step No. 12 Calculate the number of primary turns, N_p .

$$N = 1000 \cdot \sqrt{\frac{L_{(new)}}{L_{(1000)}}}$$

$$N = 1000 \cdot \sqrt{\frac{0.0216}{39.4}} = 23.4 \text{ use } 23 \text{ [turns]}$$

Step No. 13 Calculate the current density, J , using a window utilization $K_u = 0.4$.

$$J = \frac{2 \cdot (ENERGY) \cdot 10^4}{B_m \cdot A_p \cdot K_u}$$

$$J = \frac{2 \cdot 0.000333 \cdot 10^4}{0.4 \cdot 0.112 \cdot 0.4} = 372 \text{ [A/cm}^2\text{]}$$

Step No. 14 Calculate the required incremental permeability, $\Delta\mu$.

$$\Delta m = \frac{B_m \cdot MPL \cdot 10^4}{0.4 \cdot p \cdot W_a \cdot J \cdot K_u}$$

$$\Delta m = \frac{0.4 \cdot 4.7 \cdot 10^4}{0.4 \cdot 3.14 \cdot 0.95 \cdot 372 \cdot 0.4} = 106 \text{ use } 125$$

Step No. 15 Calculate the ac flux density in TI, B_{ac} , $\Delta I = I_{ppk}$.

$$B_{ac} = \frac{0.4 \cdot p \cdot N_p \cdot \frac{\Delta I}{2} \cdot \Delta m \cdot 10^{-4}}{MPL}$$

$$B_{ac} = \frac{0.4 \cdot 3.14 \cdot 23 \cdot 2.775 \cdot 125 \cdot 10^{-4}}{4.7} = 0.213 \text{ [T]}$$

Step No. 16 Calculate the primary wire area, $A_{pw(B)}$.

$$A_{pw(B)} = \frac{I_{prms}}{J}$$

$$A_{pw(B)} = \frac{2.27}{372} = 0.0061 \text{ [cm}^2\text{]}$$

Step No. 17 Calculate the required number of primary strands, S_{np} .

$$S_{np} = \frac{A_{wp(B)}}{\#26(\text{barearea})}$$

$$S_{np} = \frac{0.0061}{0.00128} = 4.77 \text{ use } 5$$

Step No. 18 Calculate the primary the new $\mu\Omega/\text{cm}$.

$$(new) m\Omega / \text{cm} = \frac{m\Omega / \text{cm}}{S_{np}}$$

$$(new) m\Omega / \text{cm} = \frac{1345}{5} = 269$$

Step No. 19 Calculate the primary winding resistance, R_p .

$$R_p = MLT \cdot N_p \cdot \frac{m\Omega}{cm} \cdot 10^{-6}$$

$$R_p = 2.8 \cdot 23 \cdot 269 \cdot 10^{-6} = 0.0173 [\Omega]$$

Step No. 20 Calculate the primary copper loss, P_p .

$$P_p = I_p^2 \cdot R_p$$

$$P_p = 2.27^2 \cdot 0.0173 = 0.0891 [\text{W}]$$

Step No. 21 Calculate the secondary turns, N_s .

$$N_s = \frac{N_p \cdot (V_o + V_d) \cdot (1 - D_{\max} - D_w)}{V_p \cdot D_{\max}}$$

$$N_s = \frac{23 \cdot (5 + 1) \cdot (1 - 0.5 - 0.1)}{24 \cdot 0.5} = 6.9 \text{ use } 7 [\text{turns}]$$

Step No. 22 Calculate the secondary peak current, I_{spk} .

$$I_{spk} = \frac{2 \cdot I_o}{1 - D_{\max} - D_w}$$

$$I_{spk} = \frac{2 \cdot 5}{1 - 0.5 - 0.1} = 16.7 [\text{A}]$$

Step No. 23 Calculate the secondary rms current, I_{srms} .

$$I_{srms} = I_{spk} \cdot \sqrt{\frac{1 - D_{\max} - D_w}{3}}$$

$$I_{srms} = 16.7 \cdot \sqrt{\frac{1 - 0.5 - 0.1}{3}} = 7.47 [\text{A}]$$

Step No. 24 Calculate the secondary wire area, $A_{sw(B)}$.

$$A_{sw(B)} = \frac{I_{srms}}{J}$$

$$A_{sw(B)} = \frac{7.47}{372} = 0.0201 [\text{cm}^2]$$

Step No. 25 Calculate the required number of secondary strands, S_{ns} .

$$S_{ns} = \frac{A_{ws}}{\text{wire}_A}$$

$$S_{ns} = \frac{0.0201}{0.00128} = 15.7 \text{ use } 16$$

Step No. 26 Calculate the secondary the $\mu\Omega/\text{cm}$.

$$(\text{new}) \mu\Omega/\text{cm} = \frac{m\Omega/\text{cm}}{S_{ns}}$$

$$(\text{new}) \mu\Omega/\text{cm} = \frac{1345}{16} = 84$$

Step No. 27 Calculate the winding resistance, R_s .

$$R_s = MLT \cdot N_s \cdot \frac{m\Omega}{cm} \cdot 10^{-6}$$

$$R_s = 2.8 \cdot 7 \cdot 84 \cdot 10^{-6} = 0.00165 [\Omega]$$

Step No. 28 Calculate the secondary copper loss, P_s .

$$P_s = I_s^2 \cdot R_s$$

$$P_s = 7.47^2 \cdot 0.00165 = 0.0921 \text{ [W]}$$

Step No. 29 Calculate the window utilization, K_u .

$$[turns] = N_p \cdot S_{np} \text{ [primary]}$$

$$[turns] = 23 \cdot 7 = 161 \text{ [primary]}$$

$$[turns] = N_s \cdot S_{ns} \text{ [secondary]}$$

$$[turns] = 7 \cdot 16 = 112 \text{ [secondary]}$$

$$N_t = 273 \text{ turns \# 26}$$

$$K_u = \frac{N_t \cdot A_w}{W_a}$$

$$K_u = \frac{273 \cdot 0.00128}{0.95} = 0.368$$

Step No. 30 Calculate the total copper loss, P_{cu} .

$$P_{cu} = P_p + P_s$$

$$P_{cu} = 0.0891 + 0.0921 = 0.181 \text{ [W]}$$

Step No. 31 Calculate the regulation, α , for this design.

$$a = \frac{P_{cu}}{P_o} \cdot 100\%$$

$$a = \frac{0.181}{30} \cdot 100 = 0.603 \text{ [%]}$$

Step No. 32 Calculate the magnetizing force in Oersteds, H .

$$H = \frac{0.4 \cdot \pi \cdot N_p \cdot I_{pk}}{MPL}$$

$$H = \frac{0.4 \cdot 3.14 \cdot 23 \cdot 5.55}{4.7} = 34.1 \text{ [Oe]}$$

Step No. 33 Calculate the Watts per kilogram, WK .

$$W/K = 8.64 \cdot 10^{-7} \cdot f^{1.28} \cdot B_{ac}^{2.112}$$

$$W/K = 8.64 \cdot 10^{-7} \cdot 100,000^{1.28} \cdot 0.213^{2.112} = 48.5 \text{ [W/kg] or [mW/g]}$$

Step No. 34 Calculate the core loss, P_{fe} .

$$P_{fe} = \frac{mW}{g} \cdot W_{fe} \cdot 10^3$$

$$P_{fe} = 48.5 \cdot 4.1 \cdot 10^{-3} = 0.199 \text{ [W]}$$

Step No. 35 Calculate the total loss, core P_{fe} and copper P_{cu} , in watts P_Σ .

$$P_\Sigma = P_{fe} + P_{cu}$$

$$P_\Sigma = 0.119 + 0.181 = 0.38 \text{ [W]}$$

Step No. 36 Calculate the watt density, Ψ .

$$\Psi = \frac{P_\Sigma}{A_t}$$

$$\Psi = \frac{0.38}{19.8} = 0.0192 \text{ [W/cm}^2\text{]}$$

Step No. 37 Calculate the temperature rise T_r , in degrees °C.

$$T_r = 450 \cdot \Psi^{0.826}$$

$$T_r = 450 \cdot 0.0259^{0.826} = 22 \text{ [}^\circ\text{C]}$$

BIBLIOGRAPHY

Colonel William T. McLyman, Transformer and Inductor Design Handbook, Second Edition, Marcel Dekker Inc., New York, 1988.

Colonel William T. McLyman, Magnetic Core Selection for Transformers and Inductors, Second Edition, Marcel Dekker Inc., 1997

Colonel William T. McLyman, Designing Magnetic Components for High Frequency, dc-dc Converters, Kg Magnetics, Inc., 1993.

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